Design and Verification of MIL-STD-1553B Remote Terminal Modules

L. Karthik, K.V. Ramana Reddy, Dr. Siva Yellampalli

Abstract— MIL-STD-1553B is the military specification defining a Digital Time Division Command/Response multiplexed data bus. The 1553 data bus is a dual-redundant, bidirectional, Manchester II bi-phase encoded data bus with high bit error reliability. All bus communications are controlled and initiated by a main Bus Controller. Remote Terminal devices attached to the bus respond to the controller's commands. This protocol is used for effective communication in military and aerospace electronic systems. A remote terminal typically consists of a Manchester encoder, Manchester decoder, Command decoder, Command legalization and Control logic blocks. A Remote Terminal must be capable of receiving, decoding valid legal commands from the bus controller and respond within a well defined time of 2 to 12 µs. In this paper design and verification of MIL-STD-1553B Remote Terminal modules namely Encoder, Decoder, Command decoder and Command legalization is described. The design has been implemented using Verilog HDL and has been taken through front end of ASIC flow (viz., functional verification, lint check, etc.).

Index Terms—MIL-STD-1553B, Digital Time Division Multiplexing, Manchester II bi-phase, Remote Terminal modules, Encoder, Decoder, Command Decoder, Command Legalisation, Verilog HDL, ASIC design, Functional Verification, Lint Check

I. INTRODUCTION

Digital Time Division Command/Response Multiplexed Data Bus is defined in MIL-STD-1553B [7]. It is a bidirectional, dual redundant, Manchester II encoded data bus with low bit error rate and a high reliability. MIL-STD-1553B can be effectively used for aerospace applications and in avionic systems because of its serial, 1 Mbps data rate, very low error rate of 1 word fault per 10 million words, dual redundant and high reliability. As a result of its highly robust architecture [9], MIL-STD-1553B is used as a means of efficient communication network in surface based launch vehicles, submarines, surface based targets and target drones, satellites and space systems including the present International Space Station.. This paper discusses a method for designing remote terminal modules of MIL-STD-1553B namely Manchester encoder, Manchester decoder, command decoder and command legalisation based on NRZ, manchester II bi-phase encoding and decoding schemes. The design blocks are simulated and verified for their functionality. Linting analyzes the HDL code and reports warnings and errors in the design as per DRC.

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II. MATERIALS AND METHODS

1. BLOCK DIAGRAM

a) ENCODER



Fig 1: Block diagram of Encoder

Fig 1 shows the block diagram of a Manchester encoder. The protocol requires the 16-bit data to be transmitted in following format: command or data sync occupying 3 bit times and Manchester II bi-phase encoded occupying 17 bit times (16 bit data and one bit of odd parity) in a 20 bit word frame as shown in Fig 2.



Fig 2: Word format of MIL-STD-1553B Encoder

Manchester encoding transitions at the center of the bit time and provide a self clocking waveform with equal positive and negative values. Logic '1' is a signal that transits from a positive level to a negative level. Logic'0' is a signal that transits from a negative level to a positive level.

Manchester encoding is shown in Fig 3. Encoder uses a 2 MHz clock. At power on, reset signal is asserted for two clock cycles. When logic '1' is received on the start input of encoder, it begins the manchester encoding of the 16 bit data.

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Fig 5: Manchester II BI-Phase Level Encoding

Table I gives the description of inputs and outputs of Encoder.

SL	Parameter	Input/	Description
NO		output	_
1	clk	input	2 MHz with a period of 250 ns
2	rst	input	1 clock cycle
3	start	input	2 clock cycles initially
4	cmd_data	input	Cmd or data sync
5	din [15:0]	input	16 bit hex data
6	man_enc	output	Serial manchester encoded data
7	Busy	output	Indicates the end of manchester encoding

Table I Input/output description of Encoder

b) DECODER



Fig 4: Block diagram of decoder

Fig 4 shows the block diagram of decoder. It reads the serial data from the 1553B data bus decodes it and verifies for valid sync and data bits. Manchester encoded data is deserialized and the 16 bit word is decoded. Error detection viz., Manchester error, parity error and bit length error are also implemented in this module. Table II gives the description of inputs and outputs of Decoder.

Table II Input/output description of Decoder

SL	Parameter	Input/	Description
NO		output	-
1	clk	input	2 MHz with a period of 250 ns
2	rst	input	1 clock cycle
3	man_enc	input	Serial manchester encoded data from encoder
4	dout [15:0]	output	16 bit hex decoded data
5	valid_data_out	output	Indicates whether decoded data is valid
6	cmd_data_out	output	Indicates cmd sync or data sync pattern
7	man_error	output	Indicates manchester error in the data bus
8	parity_error	output	Indicates parity error in the data bus
9	bit_length_error	output	Indicates bit length error in the data bus

Fig 5 shows the proposed finite state machine diagram for decoder based on the state transition tabulated in table III.



Fig 5: Proposed FSM for decoder

where, S1 is rst condition

- S2 is shifting state
- S3 is bit capture state
- S4 is cmd sync detection state
- S5 is data sync detection state
- S6 is error detection state

S7 is normal decoding state with no errors introduced

Table III State transition for decoder

SL.	PS	NS	Input
NO			
1	rst	rst	rst = 1
2	rst	shift	rst = 1
3	shift	shift	-
4	shift	cmd sync	data_bus [47:40] == 'b11100001 'b11100010
5	shift	data sync	data_bus [47:40] == 'b00011101 'b00011110
6	shift	bit capture	-
7	cmd	error	man error = dout[16:0]==~dout[16:0]
	sync		bit length error = (count>message)&(count-message<39)
			parity error = $\sim((dout[17:2])=dout[1])$
8	cmd	no error	-
	sync		
9	data	error	man error = dout[16:0]==~dout[16:0]
	sync		bit length error = (count>message)&(count-message<39)
			parity error = $\sim (^(dout[17:2])) = dout[1]$
10	data	no error	-
	sync		
11	bit	rst	-
	capture		
12	no	rst	-
	error		

c) COMMAND DECODER



Fig 6:Block diagram of command decoder

The 16 bit command consists of RT Address, Transmit or Receive, RT Sub address and Word count or Mode command as shown in Fig 6.

Command decoder decodes the command words and also determines whether the received command is a broadcast command. Fig 7 shows the format of a MIL-STD-1553B command word.



Fig 7: Format of MIL-STD-1553B Command Word

Table IV illustrates the various command word combinations along with their respective data transfer schemes.

	Table IV Command Decoding						
SL. NO.	Command Word (CW)	Description					
1	CW[10] = 0	Normal Data Transfer-Receive					
2	CW[10] = 1	Normal Data Transfer-Transmit					
3	CW[15:11] = 11111 and CW[10] = 0	Broadcast Command-Receive					
4	CW[15:11] = 11111 and CW[10] = 1	Broadcast Command-Transmit					
5	CW[10] = 0 and CW[9:5] = 11111/00000	Mode Code-Receive					
6	CW[10] = 1 and CW[9:5] = 11111/00000	Mode Code-Transmit					
7	CW[15:11] = 11111 and CW[10] = 0 and CW[9:5] = 11111/00000	Broadcast Mode Code-Receive					
8	CW[15:11] = 11111 and CW[10] = 1 and CW[9:5] = 11111/00000	Broadcast Mode Code-Transmit					

d) COMMAND LEGALIZATION



Fig 8:Block diagram of command legalization

Fig 8 shows the block diagram of command legalization module. Command Legalization block contains the logic for validating all the 1553B command words. Command word is provided as an input and the logic generates the valid or invalid output indicating the legality of the received command.

Table V shows a sample register map of command illegalisation block where all Broadcast Transmit commands are illegalised.

	Reg	15/	14/	13/	12/	11/	10/	9/	8/	7/	6/	5/	4/	3/	2/	1/	0/
	No	31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Receive Subaddress	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Transmit Subaddress	2	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)	3	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Broadcast Receive	4	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Subaddress	5	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)																	
Broadcast Transmit	6	I	I	I	Ι	I	Ι	I	Ι	I	I	I	I	I	I	I	Ι
Subaddress	7	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)																	
Receive Mode Code	8	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)	9	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Transmit Mode Code	10	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)	11	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Broadcast Receive Mode	12	0	0	0	0	0	0	0	0	0	0	0	0	0	I	0	Ι
Code	13	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
(0 to 31)																	
Broadcast Transmit	14	0	0	0	0	0	0	0	0	0	0	0	0	0	I	0	Ι
Mode Code	15	I	I	I	I	I	I	Ι	I	I	I	I	I	I	I	I	I
(0 to 31)																	

Table V Command Illegalisation register map

where I-illegal command

III. RESULTS AND DISCUSSIONS

1. SIMULATION

Simulation of remote terminal modules of MIL-STD-1553B was carried out using a RTL simulator. The simulation results of the various remote terminal modules of MIL-STD-1553B are described in detail in the following sections. Verification was carried out through testbenches.

a) ENCODER

Fig 9 illustrates the working of an encoder through a simple flowchart. Fig 10 shows the simulation result of encoder which consists of cmd sync cycle occupying 20 bit times from 0.75 μ s to 20.75 μ s and data sync cycle occupying another 20 bit times from 20.75 μ s to 40.75 μ s.

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Fig 9 Proposed flowchart for Encoder

For example A5A5h was taken as the input data. Output of the encoder consists of either cmd/data sync occupying 3 bit times, serial manchester encoded II biphase data occupying 16 bit times and odd parity bits occupying the last bit time out of a 20 bit word frame as shown in Fig 10.



Fig 10: Encoder with sync, manchester encoded data and parity

Tables VI and VII illustrate an example of manchester encoding with corresponding bit times in μ s wherein 16 bit input A5A5h is encoded.

Table VI	Example	for cmd	sync cy	vcle of	encoding
					· · · · · · · · · · · · · · · · · · ·

din	binary equivalent of	manchester encoded data	sync	odd
[15:0]	din [15:0]		(cmd)	parity
A5A5	1010 0101 1010 0101	10011001011001101001100101100110 (16 µs)	111000 (3 µs)	10 (1 μs)

Table VII Example for data sync cycle of encoding

din [15:0]	binary equivalent of din [15:0]	manchester encoded data	sync (cmd)	odd parity
A5A5	1010 0101 1010 0101	10011001011001101001100101100110	000111	10
		(16 µs)	(3 µs) f	η (1 μs)

b) DECODER

Decoder reads the serial manchester encoded data from the 1553B data bus and checks for valid sync and data bits as shown in Fig 11. When the valid signal goes high decoded data is available at the output. manchester error, parity error or bit length error, if any are also detected.



Fig 11: Decoder with decoded data and valid sync

i. Manchester error

Manchester error occurs when there is no transition present in the encoding of '1' or '0'. Error is introduced through testbench by modifying the manchester encoded data 01(for data bit '0') to 00/11 or 10(for data bit '1') to 00/11 in the serial Manchester encoded data stream. The decoder detects this as manchester error as shown in Fig 12.

For example, if the 18^{th} bit of serial encoded data 3A84 is flipped from 0 to 1, invalid manchester encoding 00 is obtained which is detected as manchester error as shown in Fig 12.

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Fig 12: Decoder with manchester error

Table VIII illustrates the introduction and detection of manchester error in the serial manchester encoded data stream.

Table VIII Detection of manchester error

SL	Manchester encoded data	Description			
NO					
1	111000 01011010 10 <u>00</u> 1001 10010101 01100101 10	18th bit is flipped from 1 to			
	cmd sync manchester error parity	0 to get an invalid			
		manchester encoding <u>00</u>			
2	dout [15:0] = 3A84h	Manchester decoded data			

ii. Parity error

Parity error is recognized by the decoder when the expected parity for the 16-bit data is not same as the parity bit received in the serial data stream. Parity error detection is tested using testbench by flipping the expected parity bit in the serial manchester encoded data stream.

For example, the expected parity of input 7C12H is '0' and the received parity is '1', then parity error output is set high as illustrated in table IX. It is shown in Fig 13.

Table IX Detection of parity error

SL	Manchester encoded data		Description
NO			
1	111000 01101010 10100101 01010110 01011001	<u>10</u>	Last two odd parity bits are
	cmd sync manchester encoding	parity	flipped to $\underline{10}$ giving rise to
		error	parity error.
2	Expected parity = ~(^(7C12h)) = ~(1) = 0		Corresponds to 01 in
			manchester encoding
3	dout [15:0] = 7C12h		Manchester decoded data



Fig 13:Decoder with parity error

iii. Bit length error

Bit length error is detected when there are missing bits or additional bits in the serial data on the bus. For example, if the last two odd parity bits of the input 95A4H after encoding are missing then bit length error and manchester error are detected as illustrated in table X and is shown in Fig 14.

Table X Detection of bit length error

	=	
SL	Manchester encoded data	Description
NO		
1	111000 01101010 10100101 01010110 01011001 -	Last two odd parity bits are
	cmd manchester encoding bit length	missing giving rise to bit
	sync error	length error.
2	dout [15:0] = 95A4h	Manchester decoded data



Fig 14:Decoder with bit length error

c) COMMAND DECODER

Fig 15 illustrates a flowchart for the working of command decoder. A 16 bit data input is decoded into various fields as per the word format of command decoder as shown in fig 16 and 17.

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Fig 15: Proposed flowchart for Command Decoder



Fig 16:Decoding the normal data transfer commands

For example, normal data transfer command 38A4h is decoded into the following fields as shown in Fig 16: RT address -00111, Broadcast -0, Tx/Rx -0, mode code -0, Subaddress -00101, word count -00100



Fig 17: Decoding Broadcast commands

For example broadcast command FA0Eh is decoded into the following fields as shown in Fig 17: RTAddress – 11111, Broadcast – 1, TX/RX – 0, mode code – 0, Subaddress – 10000, word count – 01110

d) COMMAND LEGALIZATION





Fig 18 illustrates a flowchart for the working of command legalisation. A 16 bit command is given as an input and checked for its legality, based on the command illegalization register map according to table V as shown in Fig 19 and 20.



Fig 19:Command legalisation simulation window

For example commands 3A84h and 3B06h are legal as shown in Fig 19 by the cmd legal signal which is set to 1.

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Fig 20:Detecting an Illegal command

For example commands FB82h and FC00h are illegal as shown in Fig 20 by the cmd legal signal which is set to 0.

IV. LINTING RESULTS

Linting check was carried out to detect any semantic errors in the design. Table X1 mentions the error report summary obtained using the original design.

SL.	RT	Fatal	Error	Warning	Total
NO.	module	error			errors/warnings
1	Encoder	0	1	0	1
2	Decoder	0	2	0	5
		0	2	0	
		0	1	0	
3	Command	0	10	0	18
	decoder		1	0	
			7	0	
4	Command	0	0	2	2
	legalisation				

Table XI Summary for lint check with original design

Necessary changes were made in the design to minimise the errors and they have been described in detail as follows:

a) Non-blocking statements in a combinatorial block: This error is eliminated by making use of blocking assignments.

b) Internally generated resets: This error is resolved by properly defining reset conditions in the design.

c) Unregistered outputs: This error is removed by registering the outputs with respect to clock.

Table XII Summary for lint check with changed design

SL.	RT module	Fatal	Error	Warning
NO		error		
1	Encoder	0	1	0
2	Decoder	0	1	0
3	Command decoder	0	0	0
4	Command legalisation	0	0	0

V. CONCLUSION

In this paper we have discussed a method for designing the remote terminal modules of MIL-STD-1553B namely Encoder, Decoder, Command decoder and Command legalization based on Manchester II bi-phase encoding and decoding schemes. The design blocks have been simulated and verified for their functionality. Linting analyses the HDL code and reports all the potential design rule check violation, warnings and errors which may interfere during synthesis, post synthesis simulation, static timing analysis and other stages of ASIC design flow.

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ACRONYMS AND ABBREVIATIONS

The following section provides the abbreviations of the various acronyms used in this paper.

ASIC-Application Specific Integrated Circuit BC-Bus Controller DRC-Design Rule Check FSM-Finite State Machine HDL-Hardware Description Language Mbps-Megabit per second NRZ-Non Return to Zero NS-Next State PS-Present State RT-Remote Terminal

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